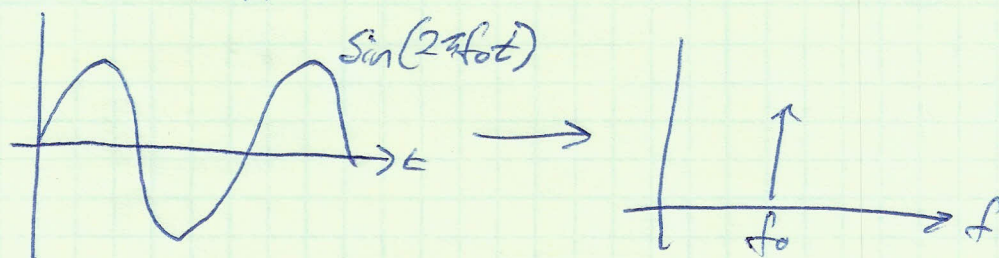
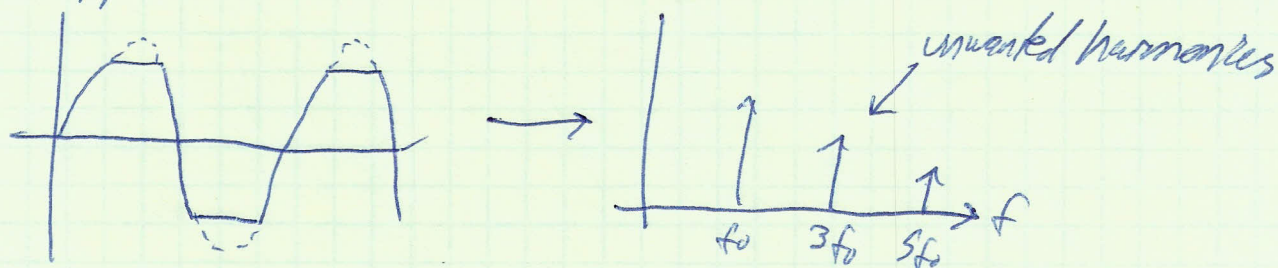


More on lab #5 - Design of an oscillator

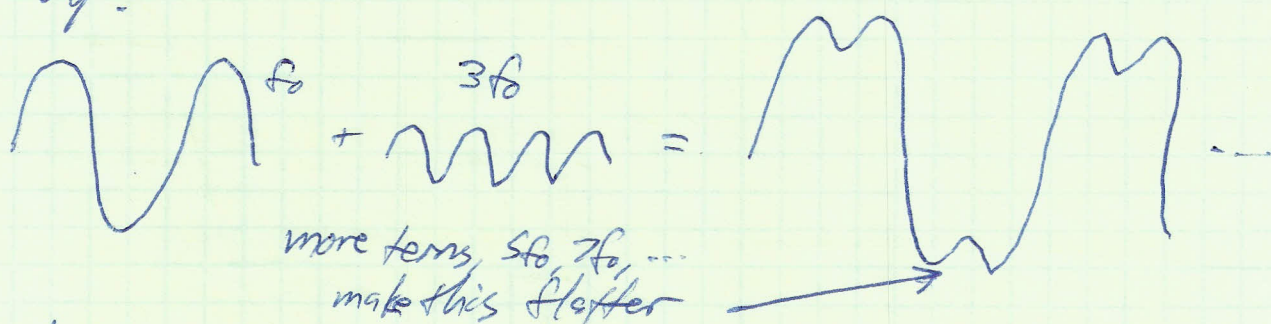
Effect of clipping



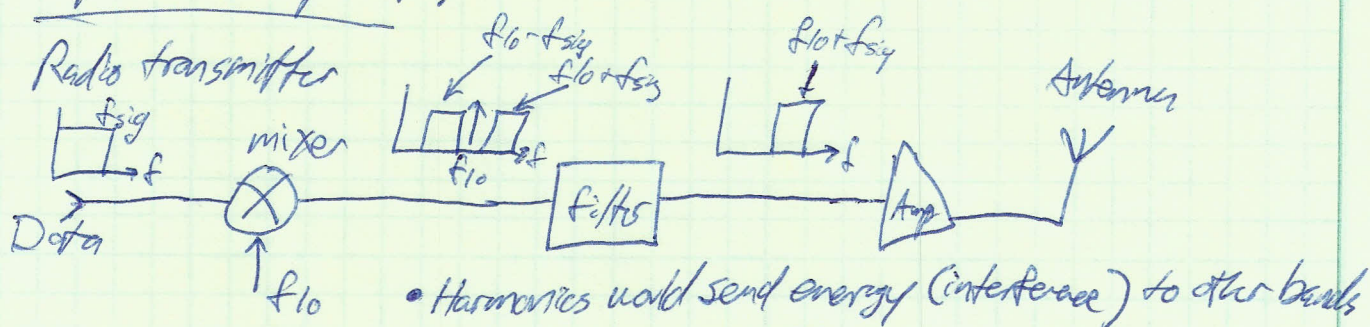
Clipped sine wave



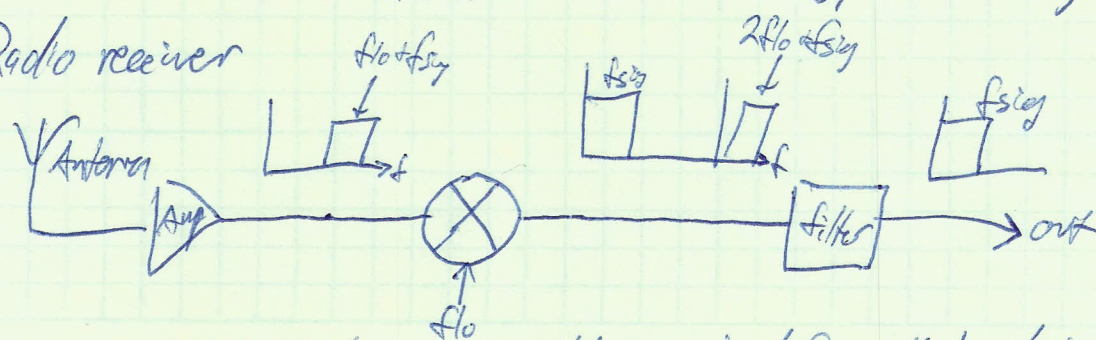
Why?



Why is this important?



Radio receiver



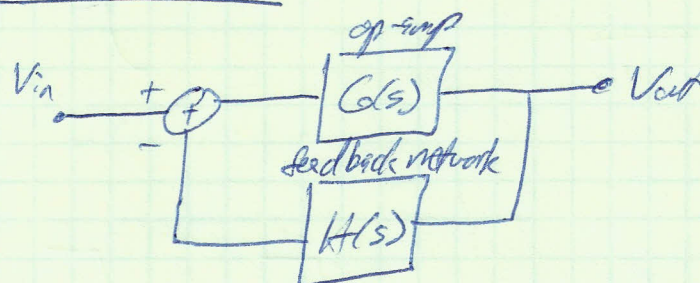
• Harmonics would cause signals from other bands to be mixed in



## Control systems

So far we have only looked at circuit feedback, but it has many applications, e.g. controlling a physical system.

### Circuit feedback

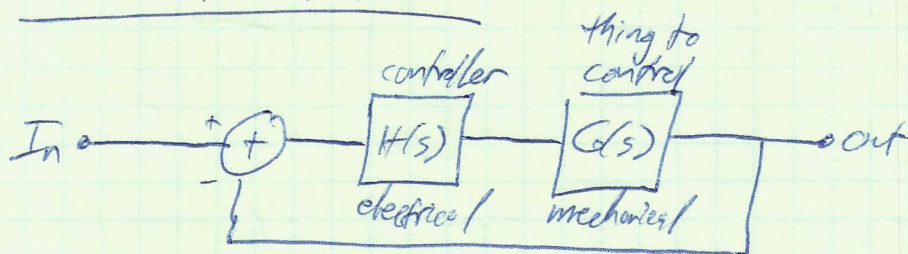


$$V_{out} = G(s)(V_{in} - H(s)V_{out})$$

$$\frac{V_{out}}{V_{in}} = \frac{G(s)}{1 + G(s)H(s)} = \frac{1}{H(s)} \cdot \frac{G(s)H(s)}{1 + G(s)H(s)}$$

$$\text{if } G(s)H(s) \gg 1, \quad \frac{V_{out}}{V_{in}} \approx \frac{1}{H(s)}$$

### Feedback control system



$$\text{out} = G(s)H(s)[In - \text{out}]$$

$$\frac{\text{out}}{In} = \frac{G(s)H(s)}{1 + G(s)H(s)}$$

In both cases, the poles of  $\frac{\text{out}}{In}$  are the same.

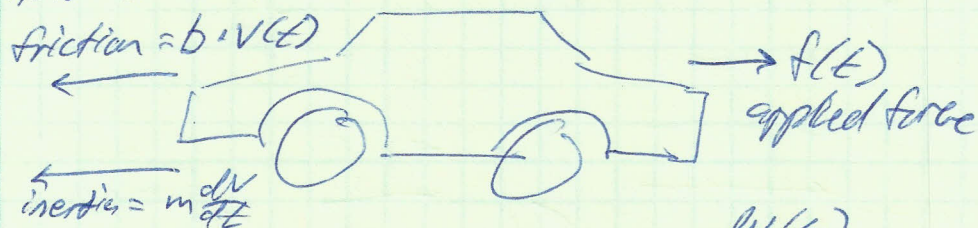
System will be stable only if there are no poles in the right half plane

What is instability for a mechanical system controller?

e.g. car cruise controller

Instability could be oscillation - constantly changing speed   $v$   
or accelerator full on   $v$

How to model a car:



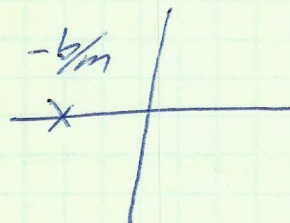
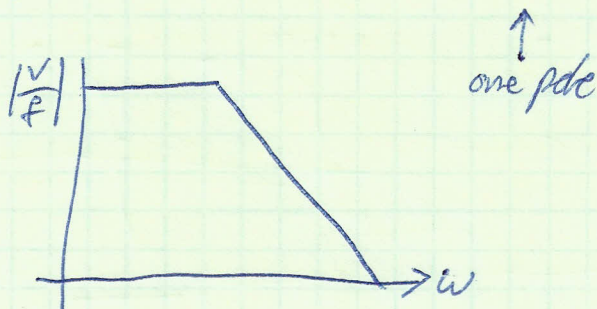
$v = \text{velocity}$

$$f(t) = b v(t) + m \frac{dv(t)}{dt}$$

laplace transform  $F(s) = b V(s) + m s V(s)$



$$\frac{V(s)}{F(s)} = \frac{1}{b + ms} = \frac{1/m}{s + b/m}$$

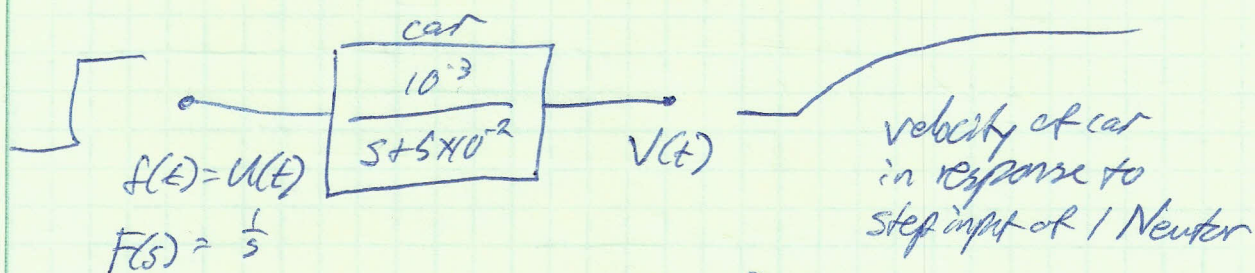


It is stable

Car behaves as a low-pass filter

example:

$$\begin{aligned} m &= 1000 \text{ kg} \\ b &= 50 \frac{\text{Ns}}{\text{m}} \end{aligned} \rightarrow \frac{V(s)}{F(s)} = \frac{10^{-3}}{s + 5 \times 10^{-2}}$$



$$V(s) = \frac{1}{s} \cdot \frac{10^{-3}}{s + 5 \times 10^{-2}} = \frac{(2 \times 10^{-2})(5 \times 10^{-2})}{s(s + 5 \times 10^{-2})}$$

$$\frac{a}{s(s+a)} \leftrightarrow U(t)[1 - e^{-at}]$$

$$V(t) = 2 \times 10^{-2} U(t) [1 - e^{-5 \times 10^{-2} t}]$$

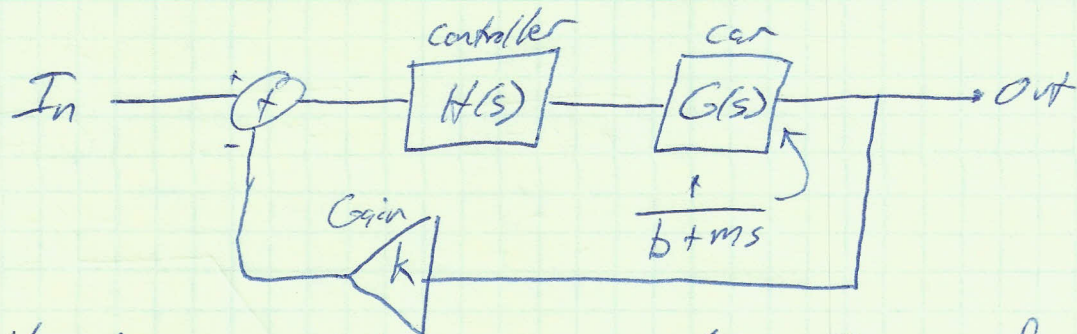
One newton force from the engine will increase speed by 0.02 m/s

How fast?  $e^{-5 \times 10^{-2} t} = 0.1$  (90% rise time)

$$t \approx 50 \text{ seconds}$$

How do we make the car reach its final speed more quickly?





How does the controller improve acceleration?

for example,  
low frequency  
response

$$Out = G(s) H(s) [In - K Out]$$

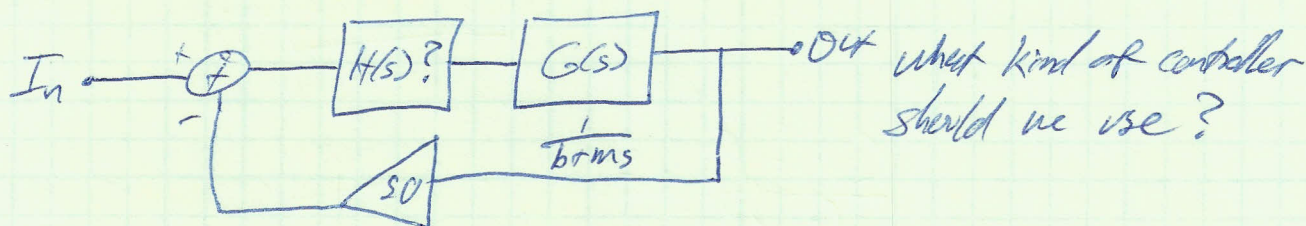
$$\frac{Out}{In} = \frac{G(s) H(s)}{1 + G(s) H(s) K} \approx \frac{1}{K} \text{ for } G(s) H(s) \gg 1$$

The controller should not change the final velocity,

Just how fast we get there.  $\rightarrow$  low frequency gain should be the same

For the car by itself, the DC gain is  $\frac{1}{b}$

The new system should have the same DC gain  $\rightarrow \frac{1}{K} = \frac{1}{b}$ ,  $K = b = 50$



what kind of controller  
should we use?

Assume a simple gain stage:  $H(s) = h$

$$\frac{Out}{In} = \frac{G(s) H(s)}{1 + G(s) H(s) K} = \frac{G(s) h}{1 + G(s) h K}$$

Poles of the transfer function:  $1 + G(s) h K = 0$

$$1 + \frac{10^{-3} h 50}{s + 5 \times 10^{-2}} = 0$$

Before feedback:

Pole at  $s = -5 \times 10^{-2}$

$$s + 5 \times 10^{-2} + 5 \times 10^{-2} h = 0$$

After feedback

Pole at  $s = -(1+h) 5 \times 10^{-2}$

$s + (1+h) 5 \times 10^{-2} = 0 \rightarrow$  pole has moved

However, the system is still stable.



Step response:

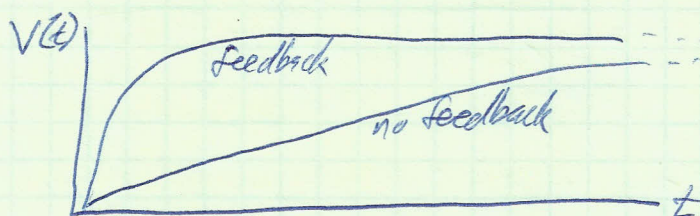
$$2 \times 10^{-2} U(t) \left[ 1 - e^{-(1+h)5 \times 10^{-2} t} \right]$$

e.g.  $h=100$

$h$  sets the step response

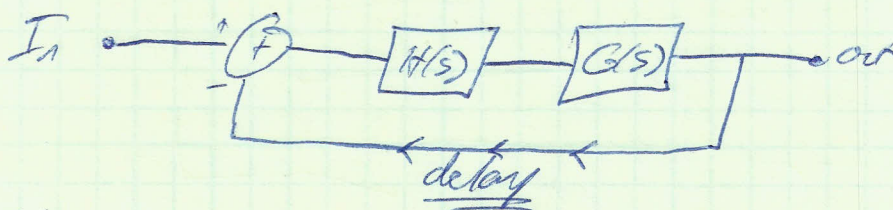
$$V(t) = 2 \times 10^{-2} U(t) [1 - e^{-5t}]$$

time to reach 90% of final velocity is  $\sim 0.5$  seconds

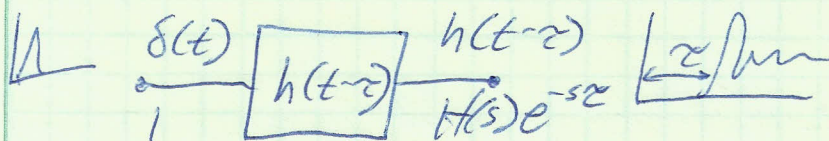
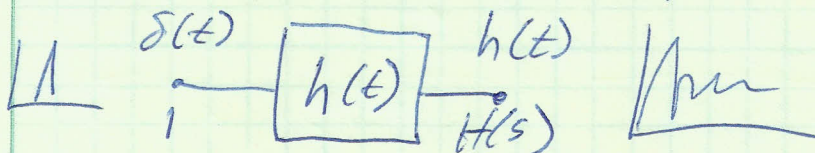


Effect of delay in feedback loop

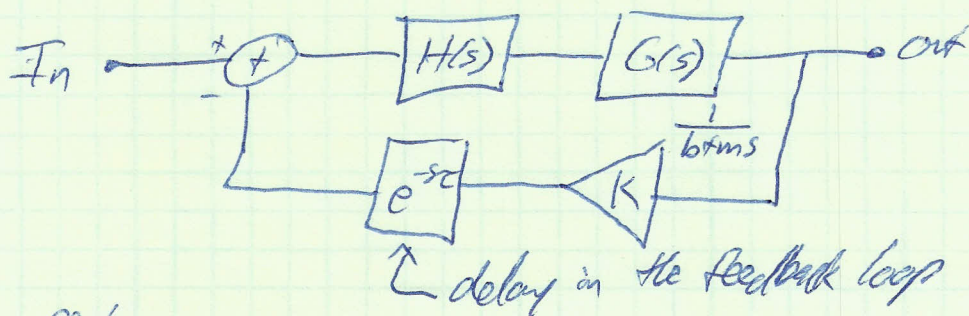
can't instantaneously correct for an error in the output



How do we model the time delay?



A delay of  $\tau$  in the time domain  
corresponds to  $e^{-s\tau}$  in the frequency domain



$$\frac{Out}{In} = \frac{G(s)H(s)}{1 + G(s)H(s)Ke^{-s\tau}}$$

$$\text{poles: } 1 + G(s)H(s)Ke^{-s\tau} = 0$$

↳ not a simple polynomial

Use polynomial approximation for  $e^{-s\tau}$

$$e^{-s\tau} \approx 1 + k_1s + k_2s^2 + k_3s^3 + \dots \quad \text{solve for } k_1, k_2, k_3, \dots$$

$$\text{Taylor series for } e^x = 1 + x + \frac{x^2}{2!} + \frac{x^3}{3!} + \dots \quad \underline{x = -s\tau}$$

$$e^{-s\tau} \approx 1 - s\tau + \frac{s^2\tau^2}{2} - \frac{s^3\tau^3}{6} + \dots$$

However,  $|e^{-s\tau}| = 1$  always.

but the Taylor series does not have constant magnitude

We need a polynomial approximation of  $e^{-s\tau}$  with magnitude = 1 for all values of  $\omega$

$$\text{Remember, } \left| \frac{a-jb}{a+jb} \right| = 1 \text{ always}$$

this is called an "all-pass" response

(for next lecture)